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Ageing effect on electromagnetic susceptibility of a phase-locked-loop

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Abstract

Phase-locked-loop in radiofrequency and mixed signal integrated circuit experience noise as electromagnetic interference coupled on input and power supply which translates to the timing jitter. Most of PLL noise analysis did not take into account the ageing effect. However device ageing can degrade the physical parameters of transistors and makes noise impact worse. This paper deals with the analyses of PLL immunity drift after accelerated ageing.

1. Introduction

During their lifetime, integrated circuits (ICs) can be affected by failure mechanisms such as electro migration, gate oxide break down or hot carrier injection [1], mainly activated by harsh environmental conditions such as high or low temperature and electrical overstress. Even if failure mechanisms do not compromise the circuit operation, IC intrinsic degradations can have a significant impact on performances, such as noise sensitivity, operating frequency... or Electromagnetic Compatibility (EMC). [2]

The concept of electromagnetic reliability (EMR) [3] has been developed to study and evaluate the impact of circuit ageing on EMC. Analog circuits as PLL are very sensitive to electromagnetic interferences (EMI) [4]. Changes of transistors physical characteristics (such as threshold voltage and mobility) and mismatches due to intrinsic degradations could affect the susceptibility of analog circuits. In this paper, we focus on the evolution of the susceptibility of a CMOS Voltage-Controlled Oscillator (VCO) to EMI coupled on the power supply after accelerated ageing, as power supply fluctuations of VCO is the dominant noise source in PLL. In the section II the principles of EMR characterization is described. EMC measurement methods and accelerated life tests are

given in section III. Section IV presents the device under test (DUT). In the fifth part, the immunity levels measurement before and after accelerated ageing test are compared and the EMI-induced failures are detailed. Finally, experimental results are summed up and discussed.

2. EMR Characterization principles

The characterization of the EMR of a circuit consists in measuring its emission and/or susceptibility level before and after an ageing procedure which accelerates intrinsic degradation mechanisms. As illustrate in Fig.1, it aims at extracting the EMC level drifts induced by ageing as accurately as possible.

Once the EMC and the ageing procedure set-ups have been completely defined, they have to be validated experimentally. The validation step must prove that ageing procedure can only have an impact on the DUT. The EMC validation step must ensure that the EMC measurements are enough repeatable to extract precise information concerning the EMC drifts related to ageing effects. Measurement repeatability and uncertainties set a limit for the consistence of EMC level drifts.

At the end of the measurement campaign, the EMC measurement results obtained before and after ageing are processed to extract statistical data concerning the EMC

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level drifts, such as the worst case and the mean drifts. These data are required to predict the risk that a component becomes in compliant after ageing and, if necessary, readjust the EMC margins at the design level.

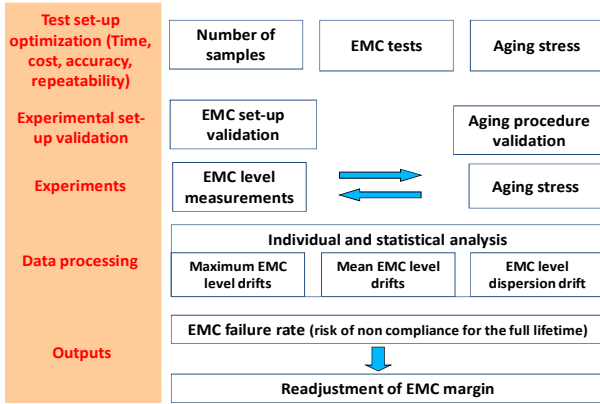


Fig.1. General description of the electromagnetic robustness of a component

3. Experimental set-up and device under test

3.1. Direct Power Injection Set-up

The direct power injection (DPI) [5] is a standard and efficient method to couple a conducted disturbance to a pin of a circuit (Fig. 2). It consists in superimposing RF disturbances to a low frequency signal through a decoupling network. Uncertainties of the method are acceptable to ensure consistency of EMR results.

Susceptibility is characterized over the band 10 MHz to 1 GHz, the maximum forward power is set to 45 dBm. Three susceptibility criteria are considered: a static margin of the output amplitude set to 20 % of the power supply Vdd, a dynamic margin set to 20% of the output signal period (8ns) and a power current limited to 40 times the nominal current.

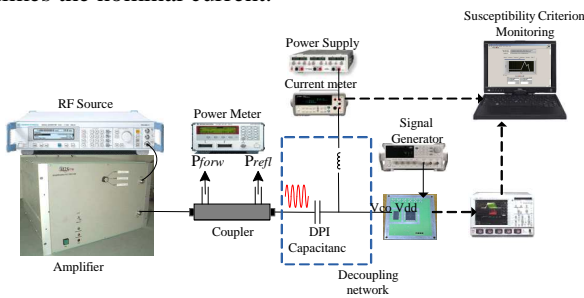


Fig. 2. DPI setup, RF injected on power supply rail of VCO

3.2. Burn-in Test Conditions

According to the standard AEC-Q100 [6], accelerated life tests are required to qualify integrated circuits and guarantee the quality and the robustness of

automotive applications. These tests consist in applying an overstress conditions during a short time in order to accelerate the damage rate for relevant degradation mechanisms. As ICs embedded in automotive applications often suffer from high temperature, High Temperature Operating Life (HTOL) which consists in applying overstresses conditions.

Ten components are tested and placed during 408 hours in a climatic chamber which regulates the temperature at 150°C +/- 2.5°C. An external source provides a power supply 10% higher than the normal supply voltage. Figure 3 describes the accelerated life test set-up. The HTOL stress mainly accelerates wear-out failure mechanisms as Time Dependent Dielectric Breakdown (TDDB) and Negative bias temperature instability (NBTI).

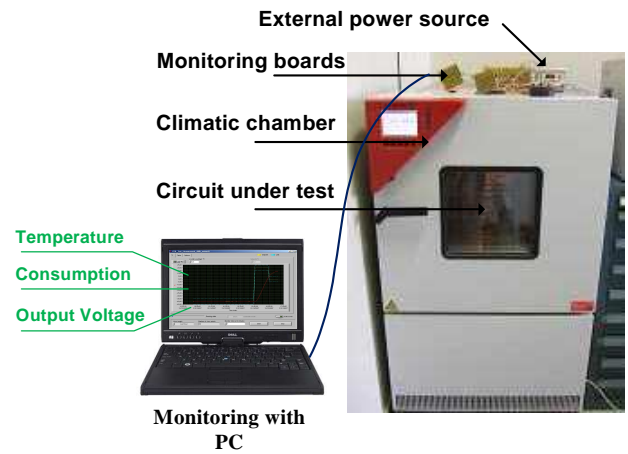


Fig. 3 Accelerated life test set up

The experimental set-up uses three boards (Tab 1). Dedicated sockets are used on the EMC and ageing boards to move the circuit under test. With this set-up, only the circuits under test are aged.

Table 1 – Three dedicated test boards

EMC board	with a RF Socket to measure susceptibility level before and after accelerated ageing
Ageing board	placed in the climatic chamber with the sockets special for ageing
Monitoring board	supply and monitor components under test during accelerated ageing

4. Device under test

The PLL is based on a delay controlled ring oscillator designed in 0.25µm SMARTMOS 8 technology from Freescale Semiconductor (Toulouse -

France), dedicated to automotive applications. The PLL is divided in three sub-blocks: VCO, fraction 1/4 frequency divider and phase comparator. The PLL operates at 24 MHz. Power supplies of these sub-blocks have been separated to identify their own sensitivities to RFI. (Fig.4)

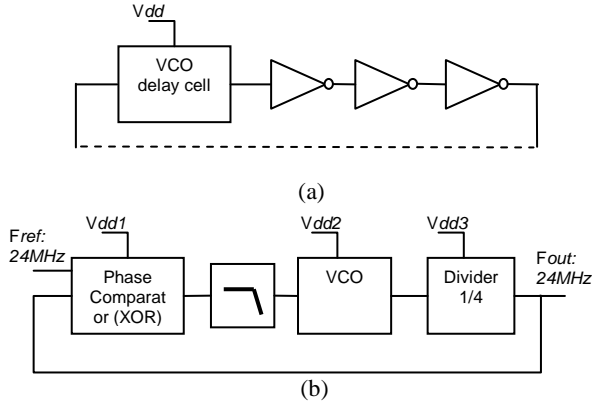


Fig. 4. (a) VCO schematic and (b) PLL schematic, it has three separated power rails for its three sub-blocks

5. Experimental results analysis

5.1 Validation of the experimental set-up

First, the repeatability of EMC test bench is evaluated for validation purpose by repeating the measurement 10 times on the same sample. The test bench repeatability is 0.27dB which is highly acceptable since the repeatability errors are inferior to the uncertainty of the DPI test bench, estimated to 2 dB.

5.2 PLL Susceptibility Analysis

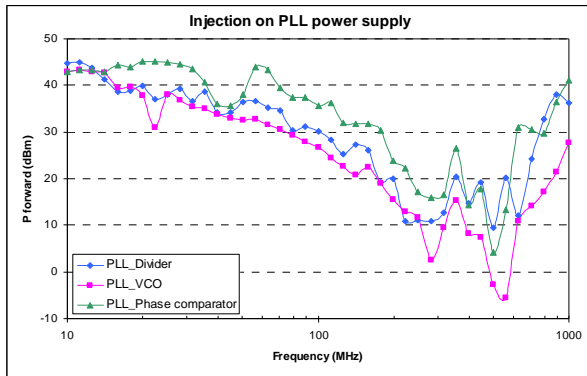


Fig. 5. PLL susceptibility threshold during conducted injection on power supply pins of the three sub-blocks (Divider, VCO, Phase comparator)

Conducted EMI injections are performed on the power supply pins of the three PLL sub-blocks. Fig. 5 presents the susceptibility threshold for injection done on three separated power rails: phase comparator, VCO,

and divider.

The comparison among three immunity curves shows that the VCO is the most sensitive block in PLL to the power supply noise.

5.3 VCO Susceptibility Analysis

VCO's output frequency generally depends on power supply voltage. The variation on power supply can change the propagation delay, thus changing the oscillation frequency in VCO. The sensitivity of VCO to the power supply voltage can be extracted by measuring the output frequency of the VCO for different power supply voltage values (from 2V to 5.5V). Fig. 6 shows the VCO output frequency as a function of its static power supply

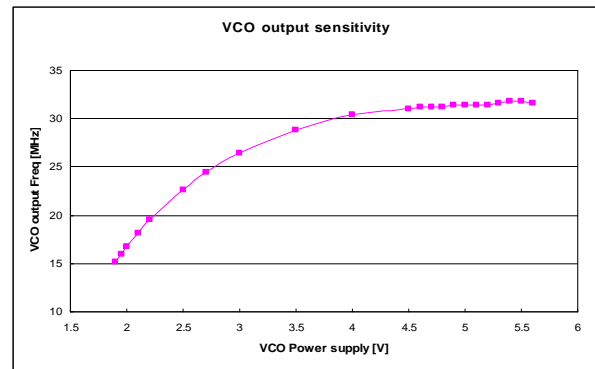


Fig. 6. Dependence of VCO output frequency to power supply voltage.

Any RFI coupled on VCO power supply can modulate the VCO output frequency at the RFI rhythm, resulting in a jitter added to the PLL output signal.

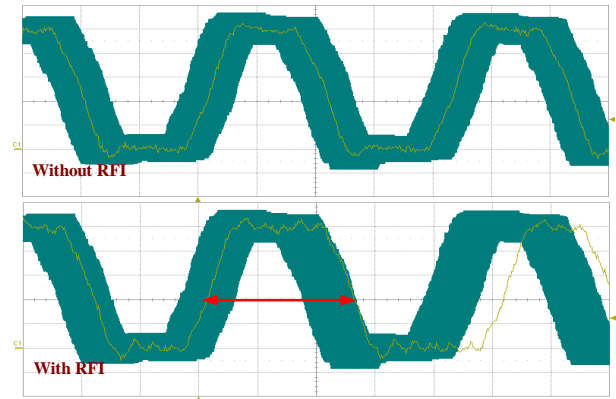


Fig. 7. PLL output signal without (on the left) and with RF injection

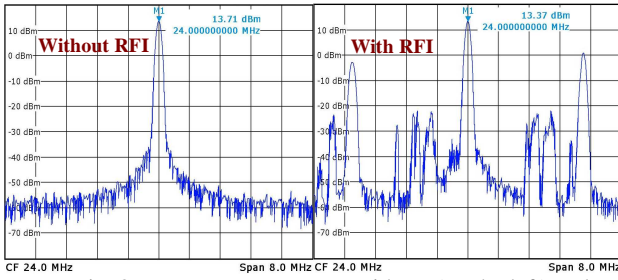


Fig. 8. PLL output spectrum without (on the left) and with RF injection

Fig. 7 shows the influence of RFI on the VCO output signal and the generation of a jitter. Fig. 8 describes the influence in frequency domain, where many spurious degrade the output signal spectral purity.

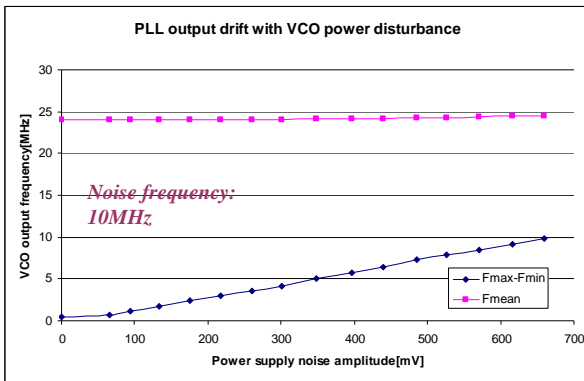
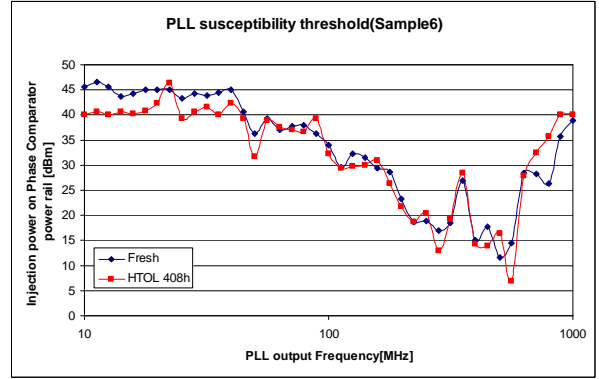


Fig. 9. PLL output spectrum without (on the left) and with RF injection

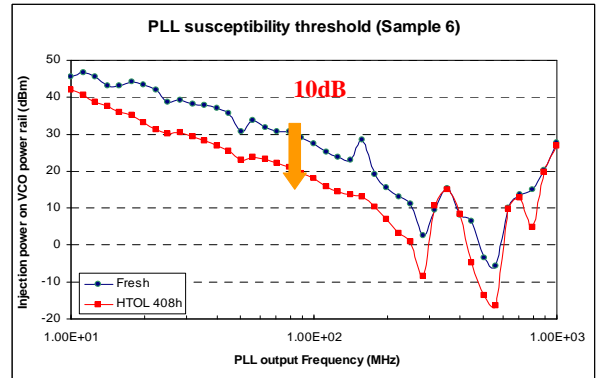
Fig. 9 plots the impact of a RFI on VCO power supply to PLL output frequency. RFI induce a considerable jitter and tends to increase the VCO output frequency.

5.2 PLL immunity Level Drift

With evaluating ageing effect on power supply immunity of three sub-blocks, the perturbation on VCO power supply exhibits a significant evolution of PLL susceptibility threshold after accelerated ageing. Fig. 9 presents the susceptibility threshold of the 6th sample for injection done on phase comparator and VCO power supply pins before and after the HTOL test. The two figures show that VCO immunity is more sensible to accelerate ageing. Its immunity level reduction can reach up to 10 dB over a large frequency range.



(a)



(b)

Fig. 9. PLL susceptibility threshold with an injection on (a) Phase comparator and (b) VCO power rails before and after ageing

A statistical analysis shows that among 10 samples after HTOL the maximum reduction is 15.88 dB; while the mean reduction for all 10 samples over all frequency is 2.57dB with a standard deviation 3.64dB. Moreover, the process dispersion among 10 samples after HTOL changes from 1.64dB to 4.83 dB.

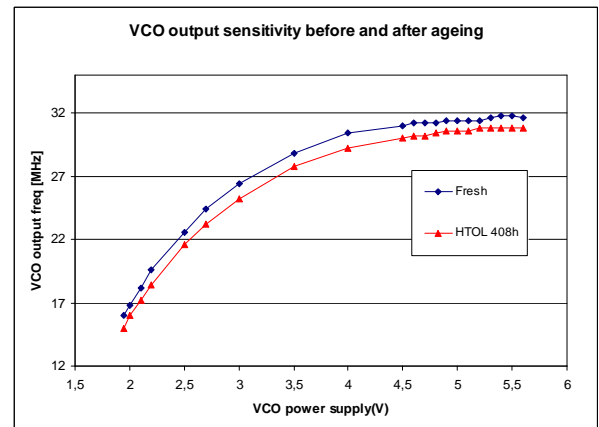


Fig.10. Dependence of VCO output frequency to power supply voltage before and after ageing

Fig 10 shows after ageing the sensitivity of the VCO has a drift about 1MHz for each VCO power supply voltage.

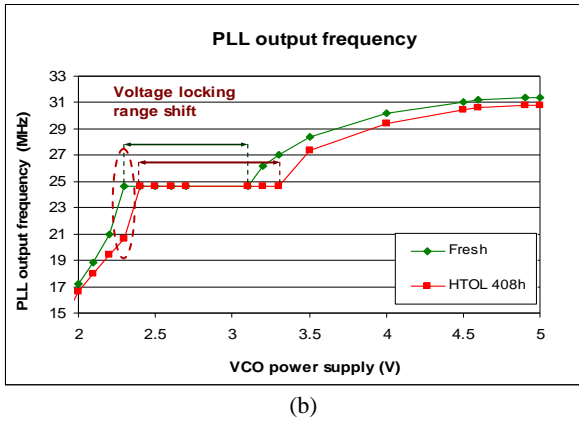


Fig.11. PLL output frequency vs. VCO power supply

This result indicates that the VCO and thus the PLL becomes more sensitive to power supply fluctuations after ageing. After ageing, a shift of the power supply voltage range ensuring the PLL locking has been measured (fig.11).

6. Discussion

After HTOL ageing tests, the functionality of PLL on all the tested samples is preserved. The current consumption and the performances are unchanged after ageing, proving the excellent robustness of the Freescale SMARTMOS 8 technology. However, accelerated-ageing test tends to decrease significantly the immunity levels of VCO and can strongly affect the failure risk of an application which uses this PLL. Moreover, the dispersions in terms of susceptibility between components tend to increase after ageing, indicating that the intrinsic degradation mechanisms follow different evolution in each sample. As HTOL mainly accelerates TDDDB and NBTI which can degrade transistor parameters, which could constitute the source of the immunity level variation after ageing. For example, the degradation of transistor transconductance (g_m) and mobility (μ) could directly affect the power supply rejection ratio and delay time of gates composing the VCO. Electrical simulations are currently leaded to confirm that the immunity reduction is induced by both a degradation of power supply noise rejection and VCO switching characteristics, provoked by changes of MOS transistor physical parameters. A model including the ageing parameters is going to be studied.

7. Conclusion

In this paper, an original study of the impact of accelerated ageing tests on PLL was presented. The results show that the functionality of circuit was preserved, but the electromagnetic susceptibility of VCO is affected by accelerated ageing test based on high temperature and electrical overvoltage conditions. The immunity level is reduced significantly over a large frequency range, which can increase interference risks with neighbor circuits. The correlation between the immunity level drift and the intrinsic degradation mechanisms needs to be studied in detail to understand the origins of the drift and predict in which extent the susceptibility threshold can decrease.

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